# MC2800SYNEVK 

## Advance Information <br> FLEX ${ }^{\text {m }}$ Paging RF/IF Frequency Synthesizer Receiver

This document contains the information of a frequency synthesized receiver prototype for FLEX roaming paging platform. This FLEX paging receiver board is a high performance RF/IF front-end for using in FLEX roaming pager. This board employs Motorola's high frequency

## Features:

- Input RF Frequency Range: 278 to 286 MHz
- Input frequency channel spacing: 25.0 kHz
- Excellent Sensitivity: - 123.0 dBm
- IF Receiver Operating Voltage: 1.15 to 1.7 V
- PLL Operating Voltage: 3.0 to 3.3 V
- Total Current Consumption:

$$
\begin{aligned}
& 4.3 \mathrm{~mA} @ \mathrm{~V} C \mathrm{C}=1.15 \mathrm{~V}, \\
& 0.08 \mathrm{~mA} @ \mathrm{~V}_{\mathrm{DD}}=3.0 \mathrm{~V}
\end{aligned}
$$

- Switchable Bit-Rate Filter to Support All Kinds of FLEX Data Rate

FLEX is a trademark of Motorola, Inc.

Figure 1. System Block Diagram


## Freescale Semiconductor, Inc. ARCHIVE INFORMATION



Figure 2. The Schematic of the Frequency Synthesized RF/IF Receiver Board


RECOMMENDED OPERATING CONDITIONS

| Characteristic | Symbol | Min | Typ | Max | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Operating Voltage | $\mathrm{V}_{\mathrm{CC}}$ | 1.15 | - | 1.7 | V |
| RF Input Frequency Range |  | 278 | - | 286 | MHz |
| PLL \& FLEX Decoder Interface Voltage | $\mathrm{V}_{\mathrm{DD}}$ | 3.0 | - | 3.3 | V |

## TEST CONDITIONS

| Characteristic | Symbol | Min | Typ | Max | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Operating Voltage | $\mathrm{V}_{\mathrm{CC}}$ | - | 1.15 | - | V |
| PLL \& FLEX Decoder Interface Voltage | $\mathrm{V}_{\mathrm{DD}}$ | - | 3.3 | - | V |
| MC2800 Regulated Output Voltage | $\mathrm{V}_{\text {reg }}$ | - | 0.980 | - | V |
| RF Input Frequency | $\mathrm{f}_{\mathrm{RF}}$ | - | 282.0000 | - | MHz |
| Synthesizer Frequency | $\mathrm{f}_{\text {syn }}$ | - | 260.6000 | - | MHz |

ELECTRICAL CHARACTERISTICS

| Characteristic | Conditions | Symbol | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| LOW NOISE AMPLIFIER |  |  |  |  |  |  |
| Power Gain [Note 1] | @ 282.000 MHz | Gp | - | 19.9 | - | dB |
| S11 [Note 1] |  | $\mathrm{S}_{11}$ | - | -10.1 | - | dB |
| Zout [Note 1] |  | $\mathrm{Z}_{\text {out }}$ | - | $227+\mathrm{j} 46$ | - | $\Omega$ |
| Noise Figure [Note 1] |  | NF | - | 3.3 | - | dB |
| Current Consumption | $\mathrm{V}_{\text {reg }}=0.980 \mathrm{~V}$ | I | - | 850 | - | $\mu \mathrm{A}$ |

SAW FILTER

| Insertion Loss | Terminating Impedance <br> $=230 \Omega$ | - | 1.0 | - | dB |
| :--- | :---: | :--- | :--- | :--- | :--- | :--- |

VOLTAGE CONTROLLED OSCILLATOR + FREQUENCY MULTIPLIER

| Desired Output Tone Frequency | 3rd Harmonic | flo1 | - | 260.6008 | - | MHz |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Desired Output Tone Amplitude |  | VLO | - | 53 | - | mV |
| 2nd Harmonic Output Power | Referenced to 3rd Harmonic | H2 | - | -46 | - | dBc |
| 4th Harmonic Output Power |  | H4 | - | -31 | - | dBc |
| Current Consumption | $\mathrm{V}_{\text {reg }}=0.980 \mathrm{~V}$ | I | - | 1050 | - | $\mu \mathrm{A}$ |

## FREQUENCY SYNTHESIZER

| Reference Frequency |  | $\mathrm{f}_{\text {ref }}$ | - | 20.95008 | - | MHz |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Reference Counter (R-Counter) |  | - | - | 09D2(H) | - |  |
| Channel Spacing |  | $\mathrm{f}_{\mathrm{ch}}$ | - | 25.000 | - | kHz |
| N-Counter Range |  | N | 2810(H) | - | 2950(H) |  |
| Synthesizer Frequency Range |  | ${ }_{\text {f }}$ gn | 256.600 | - | 264.600 | MHz |
| Phase Noise @ 10kHz Offset | $\begin{gathered} \mathrm{N}=28 \mathrm{BO}(\mathrm{H}) \\ \text { fsyn }=260.600 \mathrm{MHz} \end{gathered}$ | - | - | -95 | - | $\mathrm{dBc} / \mathrm{Hz}$ |
| $\mathrm{V}_{\text {CC }}$ Current Consumption of PLL | $\mathrm{V}_{\text {reg }}=0.980 \mathrm{~V}$ | I | - | 750 | - | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\text {DD }}$ Current Consumption of PLL | $\mathrm{V}_{\mathrm{DD}}=3.3 \mathrm{~V}$ | 1 | - | 70 | - | $\mu \mathrm{A}$ |

NOTES: 1. Output of the LNA is matched to $230 \Omega$ by inserting a $180 \Omega$ resistor in series with its output port.
2. Output of the Mixer is matched to $1.5050 \mathrm{k} \Omega$ by inserting a $1500 \Omega$ resistor in series with its output port.
3. The circuit of this test board is the same as the schematic in Figure 2.

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ELECTRICAL CHARACTERISTICS (continued)

| Characteristic | Conditions | Symbol | Min | Typ | Max | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| LNA + SAW FILTER + MIXER |  |  |  |  |  |  |
| Conversion Gain [Note 2] | Input @ 282.000 MHz <br> Output @ 21.400 MHz | GC | - | 28.2 | - | dB |
| Zout | $@ 21.400 \mathrm{MHz}$ | $\mathrm{Z}_{\text {out }}$ | - | $1620-\mathrm{j} 51$ | - | $\Omega$ |
| Current Consumption of Mixer | $\mathrm{V}_{\text {reg }}=0.980 \mathrm{~V}$ | I | - | 370 | - | $\mu \mathrm{A}$ |


| CRYSTAL FILTER |
| :--- |
| Insertion Loss |

IF RECEIVER MC2800

| Input Impedance | $@ 21.400 \mathrm{MHz}$ | $\mathrm{Z}_{\text {in }}$ | - | $1090-\mathrm{j} 140$ | - | $\Omega$ |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Second Local Oscillator Frequency |  | $\mathrm{f}_{\mathrm{LO} 2}$ | - | 20.95008 | - | MHz |
| Startup Time of 2nd LO | EN transits from ' $0^{\prime}$ to ' 1 ' |  | - | $<1.0$ | - | ms |
| Current Consumption | $\mathrm{V}_{\mathrm{CC}}=1.15 \mathrm{~V}$ |  | - | 1.45 | - | mA |

OVERALL RECEIVER PERFORMANCE
(a) Phase A Characteristics [Note 3]

| RF Input Frequency | $\mathrm{N}-\mathrm{Counter}=28 \mathrm{B8}(\mathrm{H})$ |  | - | 282.000 | - | MHz |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Sensitivity 6400/4 Level FSK 1600/2 Level FSK | $\begin{aligned} & \mathrm{R} 1={ }^{\prime} 0 ' \\ & \mathrm{R} 1=1 \end{aligned}$ |  | - | $\begin{aligned} & -123 \\ & -123 \end{aligned}$ | - | dBm dBm |
| Co-channel Rejection | $\begin{gathered} \text { 6400/4 Level FSK } \\ \text { R1 = '0' } \end{gathered}$ |  | - | -5.0 | - | dB |
| Adjacent Channel Rejection <br> Low Side <br> High Side |  |  | - | $\begin{aligned} & 56 \\ & 59 \end{aligned}$ | - | dB |
| Image Rejection First IF Second IF |  |  | - | $\begin{aligned} & 56 \\ & 51 \end{aligned}$ | - | dB |
| Intermodulation Rejection Low Side High Side |  |  | - | $\begin{aligned} & 52 \\ & 52 \end{aligned}$ | - | dB |
| Blocking <br> @ 1.0 MHz <br> @ 5.0 MHz |  |  | - | $\begin{aligned} & 73 \\ & 78 \end{aligned}$ | - | dB |
| Total Current Consumption <br> Power Up <br> Power Down | @ $\mathrm{V}_{\mathrm{CC}}=1.15 \mathrm{~V}$ <br> @ $\mathrm{V}_{\mathrm{CC}}=1.15 \mathrm{~V}$ |  | - | $\begin{gathered} 4.3 \\ 10 \end{gathered}$ | - | $\begin{aligned} & \mathrm{mA} \\ & \mu \mathrm{~A} \end{aligned}$ |

NOTES: 1. Output of the LNA is matched to $230 \Omega$ by inserting a $180 \Omega$ resistor in series with its output port.
2. Output of the Mixer is matched to $1.5050 \mathrm{k} \Omega$ by inserting a $1500 \Omega$ resistor in series with its output port.
3. The circuit of this test board is the same as the schematic in Figure 2.

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ELECTRICAL CHARACTERISTICS (continued)

| Characteristic | Conditions | Symbol | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| (b) Phase B Characteristics [Note 3] |  |  |  |  |  |  |
| RF Input Frequency | $\mathrm{N}-\mathrm{Counter}=28 \mathrm{B8}(\mathrm{H})$ |  | - | 282.000 | - | MHz |
| Sensitivity <br> 6400/4 Level FSK <br> 1600/2 Level FSK | $\begin{aligned} & \mathrm{R} 1={ }^{\prime} 0 ' \\ & \mathrm{R} 1=1 \end{aligned}$ |  | - | $\begin{aligned} & -119 \\ & -119 \end{aligned}$ | - | $\begin{aligned} & \mathrm{dBm} \\ & \mathrm{dBm} \end{aligned}$ |
| Co-channel Rejection |  |  | - | -10 | - | dB |
| Adjacent Channel Rejection <br> Low Side <br> High Side |  |  | $\begin{aligned} & - \\ & - \end{aligned}$ | $\begin{aligned} & 56 \\ & 56 \end{aligned}$ | $\begin{aligned} & - \\ & - \end{aligned}$ | dB |
| Image Rejection First IF Second IF | $\begin{gathered} \text { 6400/4 Level FSK } \\ \text { R1 = '0' } \end{gathered}$ |  | - | $\begin{aligned} & 53 \\ & 50 \end{aligned}$ | - | dB |
| Intermodulation Rejection Low Side High Side |  |  | $-$ | $\begin{aligned} & 47 \\ & 48 \end{aligned}$ | $\begin{aligned} & - \\ & - \end{aligned}$ | dB |
| Blocking <br> @ 1.0 MHz <br> @ 5.0 MHz |  |  | - | $\begin{aligned} & 69 \\ & 67 \end{aligned}$ | - | dB |
| Total Current Consumption <br> Power Up <br> Power Down | @ $\mathrm{V}_{\mathrm{CC}}=1.15 \mathrm{~V}$ <br> @ $\mathrm{V}_{\mathrm{CC}}=1.15 \mathrm{~V}$ |  | - | $\begin{gathered} 4.3 \\ 10 \end{gathered}$ | - | $\begin{aligned} & \mathrm{mA} \\ & \mu \mathrm{~A} \end{aligned}$ |

NOTES: 1. Output of the LNA is matched to $230 \Omega$ by inserting a $180 \Omega$ resistor in series with its output port.
2. Output of the Mixer is matched to $1.5050 \mathrm{k} \Omega$ by inserting a $1500 \Omega$ resistor in series with its output port.
3. The circuit of this test board is the same as the schematic in Figure 2.

PIN FUNCTION DESCRIPTION

| Pin | Symbol |  |
| :---: | :--- | :--- |
| 1 | PLL_DATA | PLL serial data input. |
| 2 | PLL_CLOCK | PLL clock signal. |
| 3 | R1 | Control bit of the MC2800 bit-rate filter. Default value $=$ ' '1'. |
| 4 | RESET | Pre-charge and reset of the MC2800. Default value $=$ '0'. |
| 5 | VDD | Digital part of PLL and FLEX decoder operating voltage. |
| 6 | VCC | Battery supplies of MC2800. |
| 7 | ENABLE | MC2800 \& PLL enable pin. (Vdd when 'ON') |
| 8 | D2 | 2-bit digital outputs of the MC2800. |
| 9 | D1 | 2-bit digital outputs of the MC2800. |
| 10 | PLL_LATCH | PLL input data latch. |
| 11 | Gnd | Ground pin. |

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## APPLICATION NOTES

This frequency synthesized FLEX paging receiver board uses Motorola's NPN silicon low-noise high-frequency transistors MMBR941LT1 to construct the RF functional blocks, including a LNA, a voltage controlled oscillator (VCO), a frequency multiplier and a mixer. NPC's PLL IC SM5166AV is used as the core of the fractional-N synthesizer. The varactor diode MMBV609LT1 is used in the VCO. The M-ary FSK IF receiver MC2800 performs the second down-conversion and demodulates the received FSK signal into two digital output bits for the baseband. Two general purpose NPN \& PNP BJTs, MMBT3904LT1 \& MMBT3906LT1 form the RSSI to RF AGC conversion circuit.

The LNA employs cascode architecture (Q1 and Q2) with its output being matched to 230 , the input impedance of the $281-\mathrm{MHz}$ SAW filter. The gain of the LNA is controlled by the base voltage of Q2, which is a function of the RSSI output of MC2800. The input matching network should be changed when an antenna precedes the LNA. Special care must be taken in characterizing the LNA and antenna to achieve a good interface match.

The frequency synthesizer comprises a VCO, a NPC's PLL IC SM5166AV, a loop filter and a frequency multiplier. The VCO (Q4) uses a modified Colpitts structure with a varactor diode D2 for the frequency tuning. C52 can be adjusted to change the frequency tuning range. The PLL uses an input reference frequency of 20.950 MHz with a divider value of 2514 (decimal), which results in a frequency spacing of 8.333 kHz . The pre-scalar value of the PLL has a control range from 10256 to 10576 (decimal). The loop filter consisting of R26, R27, C64 \& C65 has a bandwidth of about $1 / 10$ to $1 / 15$ of the frequency spacing. Q5 is used for frequency multiplication with a ratio of 3 . The third harmonic is extracted by the bandpass filter formed by C16, C17, C62, C63 \& L11. This results in a synthesized frequency range from 256.600 to 264.600 MHz . For the details of the control of PLL IC, please refer to the data sheet of SM5166AV.

A common emitter circuit (Q3) is used as the mixer with both LO and RF being ac-coupled to its base. Its output is dc-coupled to the $21.4-\mathrm{MHz}$ crystal filter. To minimize the distortion of the downconverted frequency spectrum, the filter output must be matched to MIX_IN of the MC2800.

For the characteristics of the MC2800 as well as its control (ENABLE \& RESET pins) and bit-rate filter (R1 pins) setup, please refer to the MC2800 data sheet. It is recommended that a single ceramic filter configuration be used. That is, a $450-\mathrm{kHz}$ filter (FILTER3 in Figure 2) is inserted between the MIX_OUT and IF1_IN, and a 1.0 F multi-layer ceramic chip capacitor (C73 in Figure 2) is connected between the IF1_OUT and IF2_IN. Although it is noted in the data sheet of MC2800 that 3.0 to 4.0 dB performance degradation is observed in the single ceramic filter application, this performance degradation does not occur in the overall RF/IF system performance. This is because the high gain and the low NF of the RF front-end have effectively reduced the noise contribution of the MC2800 to the overall system.

## FDB Interface

This pager board uses an 11-pin surface mount socket to interface with the pager baseband board. This interface provides 2 digital output pins (D1 \& D2), 6 control pins (ENABLE, RESET, R1, PLL_DATA, PLL_CLOCK \& PLL_LATCH) and 3 supply pins (VCC, VDD \& GND). The pin descriptions are summarized in the following table and the footprint of this socket is depicted in Figure 3. In order to provide default states for these input pins, a pull up resistor (R16) is used at R1 input pin whereas pull down resistors (R17, R20, R21, R22 \& R25) are connected at the rest of input pins.

Note: For general information regarding FLEX products, please contact the local Motorola SPS sales office or the web at http://www.motorola.com/wireless-semi.

Figure 3. The Footprint of the 11 Pin Interface Socket


54321

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Figure 4. The PCB Layout of the Receiver Board (Top Layer)


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Figure 5. The PCB Layout of the Receiver Board (Bottom Layer)


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Table 1. Component List

| C1 | 3.0 pF | L3 | 39 nH |
| :---: | :---: | :---: | :---: |
| C2, C17, C26, C27 | 4.0 pF | L4 | $2.2 \mu \mathrm{H}$ |
| C3 | 7.0 pF | L5 | $5.6 \mu \mathrm{H}$ |
| $\begin{aligned} & \text { C4, C19, C23, C53, C57, } \\ & \text { C58, C59, C60, C66, C70 } \end{aligned}$ | 1.0 nF | L6 | 100 nH |
| C5, C10, C15, C25 | 10 nF | L7 | 820 nH |
| $\begin{aligned} & \text { C6, C8, C13, C24, C35, C36, } \\ & \text { C37, C38, C40, C42, C61 } \end{aligned}$ | 100 nF | L9 | $4.7 \mu \mathrm{H}$ |
| C7, C49 | 100 pF | L10 | $2.7 \mu \mathrm{H}$ |
| C9 | 2.0 pF | L11 | 22 nH |
| C11, C52 | 33 pF | L12, L13 | BLM11B750S |
| C41 | 18 pF | Q1, Q2, Q3, Q4, Q5 | MMBR941LT1 |
| C12, C20, C47, C54 | 22 pF | Q6 | MMBT3904LT1 |
| C16, C45, D1 | Not Used | Q7 | MMBT3906LT1 |
| C18, C62 | 12 pF | R1 | 6.8 k |
| C21 | 0.5 pF | R2 | 2.2 k |
| C22 | 10 pF | R3 | 2.4 k |
| C28 | 47 nF | R4 | 200 |
| C29, C31, C44 | $4.7 \mu \mathrm{~F}$ | R5 | 430 |
| C30 | $0.1 \mu \mathrm{~F}$ | R6 | 510 |
| C32 | 2.2 nF | R7 | 10 |
| C33 | 22 nF | R8 | 8.2 k |
| C34 | 3.3 nF | R9 | 820 |
| C39 | $10 \mu \mathrm{~F}$ | R10 | 7.5 k |
| C48 C50 C67 C69 | $1.0 \mu \mathrm{~F}$ | R11, R12 | 10 M |
| C51, C63 | 1.0 pF | R13, R14, R15, R16, R17, R20, R21, R22, R25 | 100 k |
| C55 | 15 pF | R18 | 36 k |
| C56 | 56 pF | R19 | 1.6 k |
| C64 | $0.22 \mu \mathrm{~F}$ | R23 | 24 k |
| C65 | 33 nF | R24 | 9.1 k |
| C68 | $2.2 \mu \mathrm{~F}$ | R26 | 620 |
| D2 | MMBV609LT1 | R27 | 1.1 k |
| FILTER1 | SAFC281ME75N | R28 | 910 |
| FILTER 2 | MXF21.4-15A | R29 | 1.5 k |
| FILTER3 | CFWC450F | U1 | MC2800 |
| L1, L8 | 27 nH | U2 | SM5166AV |
| L2 | 33 nH | X1 | 20.945000 MHz |

Table 2. SAW Filter Specification

| Part Number | SAFC281ME75N [Note] |
| :--- | :---: |
| Nominal Center Frequency, $f_{O}$ | 281.0 MHz |
| Insertion Loss |  |
| $f_{O}-100 \mathrm{MHz}$ to $f_{O}-37.5 \mathrm{MHz}$ | 48 dB min |
| $f_{O} \pm 4.0 \mathrm{MHz}$ | 4.5 dB max |
| $f_{O}+37.5 \mathrm{MHz}$ to $f_{O}+100 \mathrm{MHz}$ | 40 dB min |
| Ripple within $f_{O} \pm 4.0 \mathrm{MHz} / \mathrm{dB}$ | 2.0 max |
| Nominal Input / Output impedance | $230 \Omega / /-0.2 \mathrm{pF}$ |

NOTE: SAW filter SAFC281ME75N is the product of muRata Manufacturing Co., Ltd. For further enquiry, please refer to the muRata's product catalog.

Table 3. Crystal Filter Specification

| Part Number | MXF21.4-15A [Note] |
| :--- | :---: |
| Nominal Center Frequency, $f_{O}$ | $21.400000 \mathrm{MHz} \pm 1.0 \mathrm{kHz}$ |
| Number of Pole | 2 poles |
| Pass Band Width | $-3.0 \mathrm{~dB} / \pm 7.5 \mathrm{kHz} \mathrm{min}$ |
| Stop Band Width | $-18 \mathrm{~dB} / \pm 25.0 \mathrm{kHz}$ max |
| Pass Band Ripple | 0.5 dB max |
| Insertion Loss | 1.5 dB max |
| Nominal Input / Output impedance | $1.5 \mathrm{k} \Omega / / 2.0 \mathrm{pF}$ |

NOTE: Crystal filter MXF21.4-15A is the product of KINSEKI, Ltd. For further enquiry, please refer to the product catalog of Kinseki, Ltd.

Table 4. Ceramic Filter Specification

| Part Number | CFWC450F [Note] |
| :--- | :---: |
| Nominal Center Frequency, $f_{O}$ | 450 kHz |
| 6.0 dB Bandwidth from $f_{O}$ | $\pm 6.0 \mathrm{kHz}$ |
| 50 dB Bandwidth from $f_{O}$ | $\pm 12.5 \mathrm{kHz}$ |
| Stop Band Attenuation <br> Within $f_{O} \pm 100 \mathrm{kHz}$ | 50 dB min |
| Ripple within $f_{O} \pm 7.5 \mathrm{kHz}$ | 3.0 dB max |
| Nominal Input/Output impedance | $1.5 \mathrm{k} \Omega$ |

NOTE: Ceramic filter CFWC450F is the product of muRata Manufacturing Co., Ltd. For further inquiry, please refer to muRata's product catalog.

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Table 5. Quartz Crystal Specification

| Part Number | 1U0209450D30F5D0Z [Note] |
| :--- | :---: |
| Nominal Frequency, $f_{O}$ | 20.945 MHz |
| Holder Type | UM-1 |
| Resonance Mode | Fundamental |
| Operating Temperature | -30 to $80^{\circ} \mathrm{C}$ |
| Frequency Tolerance @ 25 ${ }^{\circ} \mathrm{C}$ | $\pm 20 \mathrm{ppm}$ |
| Load Capacitance | 30 pF |
| Equivalent Series Resistance, ESR | $30 \Omega \mathrm{max}$ |
| Shunt Capacitance $\mathrm{C}_{0}$ | 7.0 pF |
| Drive Level | $0.5 \mu \mathrm{~W}$ |
| Aging | $\pm 5.0 \mathrm{ppm} / \mathrm{year}$ |

NOTE: The quartz crystals are the products of Hong Kong X'TALS Limited. For further inquiry, please refer to the product catalog of Hong Kong X'TALS Limited or hkxtals@HongKongCrystal.com.


#### Abstract

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